Robust Control 2018

Richard Pates and Bo Bernhardsson

Department of Automatic Control LTH, Lund University

Design for Robust Performance

Next 3 lectures:

- \mathcal{H}_{∞} -norm specifications.
- Product stability and performance.
- Ocontroller synthesis via \mathcal{H}_{∞} -norm optimisation.

Warning

"Optimisation can expose the weaknesses in thinking which are usually compensated for by soundness of intuition."

-F. Whittle

Plan of attack:

Today's topic: Formulate design specifications as \mathcal{H}_{∞} -norm criteria.

- Check soundness of intuition.
- \mathcal{H}_{∞} -norm specifications.
- Robust Performance:
 - from open loop to closed loop
 - from closed loop to open loop?
- The gang of four stability margin.







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 \mathcal{H}_∞ is the space of bounded holomorphic functions on the open right half plane, with norm

$$||f(s)||_{\infty} = \sup_{s \in \mathbb{C}_+} |f(s)|.$$



"I have never done anything 'useful'. No discovery of mine has made, or is likely to make, directly or indirectly, for good or ill, the least difference to the amenity of the world."

For stable real rational functions:

 $||f(s)||_{\infty} = \sup_{\omega \ge 0} |f(j\omega)|.$



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For matrices of stable real rational functions:

 $||F(s)||_{\infty} = \sup_{\omega \ge 0} \overline{\sigma}(F(j\omega)).$



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What does this tell us about how systems behave?

Let y(s) = T(s)r(s), and suppose we desire good tracking.



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Impossible if $||T(s)||_{\infty}$ is small!

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...possible if $||T(s)||_{\infty}$ is small!

- Can rule out large bad behaviours.
- Doesn't mean behaviour is good...

Revisit step tracking. Desire:



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Check that

$$\|S(s)\frac{1}{s}\|_{\infty} \le .05??$$

Revisit step tracking. Desire:

But

$$(1 - S(s))\frac{1}{s} \approx \frac{1}{s}$$
$$\Rightarrow S(s)\frac{1}{s} \approx 0.$$
$$\|S(s)\frac{1}{s}\|_{\infty} = \infty$$

Revisit step tracking. Desire:

$$(1 - S(s))\frac{1}{s} \approx \frac{1}{s}$$
$$\Rightarrow S(s)\frac{1}{s} \approx 0.$$

Check that

$$\|S(s)\frac{1}{s+T}\|_{\infty} \le .05$$

Use weighting functions

Use weighting functions and the \mathcal{H}_∞ norm to capture desired performance.

General advice:

- $\overline{\sigma}(S(j\omega))$ small over desired control bandwidth, and never too large.
- $\overline{\sigma}(T(j\omega))$ small at high frequencies (robustness).
- Check all the closed loop transfer functions, step responses... \mathcal{H}_{∞} norm is just a number!

13	$\ u\ _2$	$\ u\ _{\infty}$	pow(u)
$\ y\ _{2}$	$\ G\ _{\infty}$	∞	∞
$\ y\ _{\infty}$	$\ G\ _2$	$\ G\ _1$	∞
$pow(y) \ge$	0	$\leq \ G\ _{\infty}$	$\ G\ _{\infty}$



 $z(s) = A_{11}(s) + A_{12}(s)B(s)(I - A_{22}(s)B(s))^{-1}A_{21}(s)$

 $\mathcal{F}_{l}(A,B) = A_{11}(s) + A_{12}(s)B(s)(I - A_{22}(s)B(s))^{-1}A_{21}(s)$

Find A(s) for the following:

•
$$\mathcal{F}_{l}(A, B) = P(s) + \Delta(s)$$
, where $B(s) = \Delta(s)$.
• $\mathcal{F}_{l}(A, B) = W_{1}(s)(I + P(s)C(s))^{-1}$, where $B(s) = C(s)$.
• $[P(s)]$

$$\mathcal{F}_{l}(A,B) = \begin{bmatrix} P(s) \\ I \end{bmatrix} (I - C(s)P(s))^{-1} \begin{bmatrix} -C(s) & I \end{bmatrix},$$

where $B(s) = C(s).$

Read Chapter 10 of Robust and Optimal Control:

- Summing LFTs gives an LFT.
- Multiplying LFTs gives an LFT.
- Taking LFTs of LFTs gives an LFT.

Performance and robustness requirements can be specified with the LFT.



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Why use the LFT?? Formalises the mapping from controller and uncertainty to design specifications.







For example $T = G(I + G)^{-1}$. $T_P = \begin{bmatrix} I & 0 \\ I & I \end{bmatrix} \begin{bmatrix} G \\ I \end{bmatrix}$

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$$T_P = \begin{bmatrix} I & 0 \\ I & I \end{bmatrix} \begin{bmatrix} G \\ I \end{bmatrix}$$
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$$T_P = \begin{bmatrix} I & 0 \\ I & I \end{bmatrix} \begin{bmatrix} G \\ I \end{bmatrix}$$
$$\Rightarrow G_P = \begin{bmatrix} I & 0 \\ -I & I \end{bmatrix} \begin{bmatrix} T \\ I \end{bmatrix}$$
$$\Rightarrow G = T(I - T)^{-1}.$$

 $N(s), M(s) \in \mathcal{H}_{\infty}$ are right-coprime if there exist $X(s), Y(s) \in \mathcal{H}_{\infty}$ such that

$$X(s)M(s) + Y(s)N(s) = I.$$

 $N(s), M(s) \in \mathcal{H}_{\infty}$ is a normalised right-coprime factorisation of P if:

- $P(s) = N(s)M(s)^{-1}$.
- N(s), M(s) are coprime.
- $\bullet \ N(j\omega)^*N(j\omega) + M(j\omega)^*M(j\omega) = I.$

Normalised coprime factor uncertainty:

$$G_P = \begin{bmatrix} N(s) \\ M(s) \end{bmatrix} + \begin{bmatrix} \Delta_N(s) \\ \Delta_M(s) \end{bmatrix}, \left\| \begin{bmatrix} \Delta_N(s) \\ \Delta_M(s) \end{bmatrix} \right\|_{\infty} \le \gamma.$$

Frequency by frequency:





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Gang of Four Stability Margin

Assuming stability,

$$b_{P,C} = \left\| \begin{bmatrix} P \\ I \end{bmatrix} (I - CP)^{-1} \begin{bmatrix} -C & I \end{bmatrix} \right\|_{\infty}^{-1}.$$

Note $b_{P,C} = b_{C,P}$.

Gang of Four Stability Margin

Robustness properties:

- *b*_{P,C} ≤ γ if and only robust to coprime factor uncertainty of size γ.
- Robust to additive, multiplicative, inverse additive and inverse multiplicative uncertainty.

Gang of Four Stability Margin

Performance properties

- All 8 closed loop transfer functions are bounded.
- If $\underline{\sigma}(P(j\omega))$ is big, then $S_i(j\omega), S_o(j\omega)$ are small.
- If $\overline{\sigma}(P(j\omega))$ is small, then $T_i(j\omega), T_o(j\omega)$ are small.

Put
$$\overline{P}(s) = W_1(s)P(s)W_2(s)$$
, and check $b_{\overline{P},C}!$